- The single-phase induction motor is the most frequently used motor in the world
- Most appliances, such as washing machines and refrigerators, use a single-phase induction machine
- Highly reliable and economical



Figure 1 Single-phase induction motor.

- For industrial applications, the three-phase induction motor is used to drive machines
- Figure 2 Large three-phase induction motor. (Courtesy Siemens).



Figure 3 Induction motor components.



- The motor housing consists of three parts:
 - The cylindrical middle piece that holds the stator iron core,
 - The two bell-shaped end covers holding the ball bearings.
 - This motor housing is made of cast aluminum or cast iron.
 Long screws hold the three parts together.
 - The legs at the middle section permit the attachment of the motor to a base.
 - A cooling fan is attached to the shaft at the left-hand side.
 This fan blows air over the ribbed stator frame.

Figure 4 Stator of a large induction motor. (Courtesy Siemens).



- The iron core has cylindrical shape and is laminated with slots
- The iron core on the figure has paper liner insulation placed in some of the slots.
- In a three-phase motor, the three phase windings are placed in the slots
- A single-phase motor has two windings: the main and the starting windings.
- Typically, thin enamel insulated wires are used



Figure 5 Stator iron core without windings

- A single-phase motor has two windings: the main and the starting windings
- The elements of the laminated iron core are punched from a silicon iron sheet.
- The sheet has 36 slots and 4 holes for the assembly of the iron core.



Figure 6 Single-phase stator with main windings.

- The elements of the laminated iron core are punched from a silicon iron sheet.
- The sheet has 36 slots and 4 holes for the assembly of the iron core



Figure 7 Stator iron core sheet.

Figure 8 Stator and rotor magnetic circuit



Squirrel cage rotor.

- This rotor has a laminated iron core with slots, and is mounted on a shaft.
- Aluminum bars are molded in the slots and the bars are short circuited with two end rings.
- The bars are slanted on a small rotor to reduce audible noise.
- Fins are placed on the ring that shorts the bars. These fins work as a fan and improve cooling.



Figure 9 Squirrel cage rotor concept.



Figure 10 Squirrel cage rotor.

Wound rotor.

- Most motors use the squirrel-cage rotor because of the robust and maintenance-free construction.
- However, large, older motors use a wound rotor with three phase windings placed in the rotor slots.
- The windings are connected in a three-wire wye.
- The ends of the windings are connected to three slip rings.
- Resistors or power supplies are connected to the slip rings through brushes for reduction of starting current and speed control



Figure 11 Rotor of a large induction motor. (Courtesy Siemens).

Operating principle

- This two-pole motor has three stator phase windings, connected in three-wire wye.
- Each phase has $2 \times 3 = 6$ slots. The phases are shifted by 120°
- The squirrel cage rotor has shortcircuited bars.
- The motor is supplied by balanced three-phase voltage at the terminals.
- The stator three-phase windings can also be connected in a delta configuration.



Figure 11 Connection diagram of a two-pole induction motor with squirrel cage rotor.

Operation Principle

- The three-phase stator is supplied by balanced threephase voltage that drives an ac magnetizing current through each phase winding.
- The magnetizing current in each phase generates a pulsating ac flux.
- The flux amplitude varies sinusoidally and the direction of the flux is perpendicular to the phase winding.

Operation Principle

- The three-phase stator is supplied by balanced threephase voltage that drives an ac magnetizing current through each phase winding.
- The magnetizing current in each phase generates a pulsating ac flux.
- The total flux in the machine is the sum of the three fluxes.
- The summation of the three ac fluxes results in a rotating flux, which turns with constant speed and has constant amplitude.

Operation Principle

- The rotating flux induces a voltage in the shortcircuited bars of the rotor. This voltage drives current through the bars.
- The induced voltage is proportional with the difference of motor and synchronous speed. Consequently the motor speed is less than the synchronous speed
- The interaction of the rotating flux and the rotor current generates a force that drives the motor.
- The force is proportional with the flux density and the rotor bar current

- The figure shows the three components of the magnetic field at a phase angle of -60° .
- Each phase generates a magnetic field vector.
- The vector sum of the component vectors Φ_a , Φ_b , Φ_c gives the resulting rotating field vector Φ_{rot} ,
- The amplitude is 1.5 times the individual phase vector amplitudes, and Φ_{rot} rotates with constant speed.



Figure 12 Three-phase windinggenerated rotating magnetic field.

Induced Voltage Generation

Faraday's law

- Voltage is induced in a conductor that moves perpendicular to a magnetic field,
- The induced voltage is:

> Figure 14 Voltage induced in a conductor moving through a magnetic field.

- The three-phase winding on the stator generates a rotating field.
- The rotor bar cuts the magnetic field lines as the field rotates.
- The rotating field induces a voltage in the short-circuited rotor bars
- The induced voltage is proportional to the speed difference between the rotating field and the spinning rotor



$$\mathbf{V} = \mathbf{B} \mathbf{L} \left(\mathbf{v}_{syn} - \mathbf{v}_{m} \right)$$

- The speed of flux cutting is the difference between the magnetic field speed and the rotor speed.
- The two speeds can be calculated by using the radius at the rotor bar location and the rotational speed.

$$v_{syn} = 2\pi r_{rot} n_{syn}$$

$$v_{mot} = 2\pi r_{rot} n_m \quad \mathbf{V}_{bar} = 2\pi r_{rot} \mathbf{B} \ell_{rot} (\mathbf{n}_{syn} - \mathbf{n}_m)$$



- The voltage and current generation in the rotor bar require a speed difference between the rotating field and the rotor.
- Consequently, the rotor speed is always less than the magnetic field speed.
- The relative speed difference is the *slip*, which is calculated using

$$s = \frac{n_{sy} - n_m}{n_{sy}} = \frac{\omega_{sy} - \omega_m}{\omega_{sy}}$$

The synchronous speed is



Motor Force Generation

• The interaction between the magnetic field B and the current generates a force

 $\mathbf{F} = \mathbf{B} \mathbf{L} \mathbf{I}$



Figure 15 Force direction on a currentcarrying conductor placed in a magnetic field (B) (current into the page).

Force generation in a motor

- The three-phase winding generates a rotating field;
- The rotating field induces a current in the rotor bars;
- The current generation requires a speed difference between the rotor and the magnetic field;
- The interaction between the field and the current produces the driving force.



Figure 16 Rotating magnetic field generated driving force.

Equivalent circuit

- An induction motor has two magnetically coupled circuits: the stator and the rotor. The latter is short-circuited.
- This is similar to a transformer, whose secondary is rotating and short-circuited.
- The motor has balanced three-phase circuits; consequently, the single-phase representation is sufficient.
- Both the stator and rotor have windings, which have resistance and leakage inductance.
- The stator and rotor winding are represented by a resistance and leakage reactance connected in series

- A transformer represents the magnetic coupling between the two circuits.
- The stator produces a rotating magnetic field that induces voltage in both windings.
 - A magnetizing reactance (Xm) and a resistance connected in parallel represent the magnetic field generation.
 - The resistance (Rc) represents the eddy current and hysteresis losses in the iron core
- The induced voltage is depend on the slip and the turn ratio



Figure 17 Single-phase equivalent circuit of a threephase induction motor.

- In this circuit, the magnetizing reactance generates a flux that links with both the stator and the rotor and induces a voltage in both circuits.
- The magnetic flux rotates with constant amplitude and synchronous speed.
- This flux cuts the stationary conductors of the stator with the synchronous speed and induces a 60 Hz voltage in the stator windings.
- The rms value of the voltage induced in the stator is:

$$V_{sta} = \frac{N_{sta} \, \Phi_{\max} \, \omega_{sy}}{\sqrt{2}}$$

- The flux rotates with the synchronous speed and the rotor with the motor speed.
- Consequently, the flux cuts the rotor conductors with the speed difference between the rotating flux and the rotor.
- The speed difference is calculated using the slip equation:

$$(\omega_{sy} - \omega_m) = \omega_{sy} s$$

• The induced voltage is:

$$V_{rot} = \frac{N_{rot} \Phi_{\max} (\omega_{sy} - \omega_m)}{\sqrt{2}} = \frac{N_{rot} \Phi_{\max} \omega_{sy} s}{\sqrt{2}}$$

• The division of the rotor and stator induced voltage results in: N

$$V_{rot} = \frac{N_{rot}}{N_{sta}} V_{sta} \ s = V_{rot_s} \ s$$

• This speed difference determines the frequency of the rotor current

$$f_{rot} = \frac{\omega_{rot}}{2\pi} = \frac{\omega_{sy} - \omega_m}{2\pi} = \frac{\omega_{sy} s}{2\pi} = s f_{sy}$$

• The rotor circuit leakage reactance is:

$$X_{rot_m} = L_{rot} \omega_{rot} = L_{rot} \omega_{sy} s = X_{rot} s$$
• The relation between rotor current and the rotorinduced voltage is calculated by the loop voltage equation:

$$\mathbf{V_{rot}} = \mathbf{V_{rot_s}} \ s = \mathbf{I_{rot}} \ (R_{rot} + j X_{rot} \ s)$$

• The division of this equation with the slip yields

$$\mathbf{V_{rot_s}} = \mathbf{I_{rot}} \left(\frac{R_{rot}}{s} + j X_{rot} \right)$$

• The implementation of this equation simplifies the equivalent circuit



Figure 18 Modified equivalent circuit of a three-phase induction motor.

The rotor impedance is transferred to the stator side. This eliminates the transformer



Figure 19 Simplified equivalent circuit of a three-phase induction motor.

• The last modification of the equivalent circuit is the separation of the rotor resistance into two parts:

$$\frac{R_{rot_t}}{s} = R_{rot_t} + \frac{[1-s]}{s} R_{rot_t}$$

• The obtained resistance represents the outgoing mechanical power

$$\frac{\left[1-s\right]}{s}R_{rot_t}$$



Figure 20 Final single-phase equivalent circuit of a three-phase induction motor.

Motor performance

- Figure 21 shows the energy balance in a motor.
- The supply power is:

$$P_{sup} = Re(S_{sup}) = Re(3 V_{sup} I_{sta}^{*})$$

- The power transferred through the air gap by the magnetic coupling is the input power (P_{sup}) minus the stator copper loss and the magnetizing (stator iron) loss.
- The electrically developed power (P_{dv}) is the difference between the air gap power (P_{ag}) and rotor copper loss.

• The electrically developed power can be computed from the power dissipated in the second term of rotor resistance:

$$\mathbf{P}_{\mathrm{dv}} = 3 \left| \mathbf{I}_{\mathrm{rot}_{\mathrm{t}}} \right|^{2} \left(R_{\mathrm{rot}_{\mathrm{t}}} \frac{1-s}{s} \right)$$

 The subtraction of the mechanical ventilation and friction losses (P_{mloss}) from the developed power gives the mechanical output power

$$P_{out} = P_{dv} - P_{mloss}$$

• The motor efficiency:

$$\eta = \frac{P_{out}}{P_{sup}}$$

• Motor torque:

$$M = \frac{P_{out}}{\omega_{m}}$$



Figure 21 Motor energy balance flow diagram.

7.3.4 Motor performance analysis

1) Motor impedance

$$Z_{\text{rot}_t}(s) := j \cdot X_{\text{rot}_t} + R_{\text{rot}_t} + R_{\text{rot}_t} \cdot \frac{(1-s)}{s}$$

$$Z_{\mathrm{m}} \coloneqq \frac{\mathbf{J} \cdot \mathbf{X}_{\mathrm{m}} \cdot \mathbf{R}_{\mathrm{c}}}{\mathbf{j} \cdot \mathbf{X}_{\mathrm{m}} + \mathbf{R}_{\mathrm{c}}}$$

$$Z_{sta} := j \cdot X_{sta} + R_{sta}$$

$$Z_{\text{mot}}(s) := Z_{\text{sta}} + \frac{Z_{\text{m}} \cdot Z_{\text{rot}_t}(s)}{Z_{\text{m}} + Z_{\text{rot}_t}(s)}$$



Figure 7.22 Simplified motor equivalent circuit.

2) Motor Current

$$V_{sup} := \frac{V_{mot}}{\sqrt{3}}$$

$$I_{sta}(s) := \frac{V_{sup}}{Z_{mot}(s)}$$

$$I_{rot_t}(s) := I_{sta}(s) \cdot \frac{Z_m}{Z_m + Z_{rot_t}(s)}$$



Figure 7.22 Simplified motor equivalent circuit.

3) Motor Input Power

$$S_{sup}(s) := 3 \cdot V_{sup} \cdot \overline{I_{sta}(s)} \qquad P_{sup}(s) := \text{Re}(S_{sup}(s))$$
$$Pf_{sup}(s) := \frac{P_{sup}(s)}{|S_{sup}(s)|} \qquad Q_{sup}(s) := \text{Im}(S_{sup}(s))$$

4) Motor Output Power and efficiency

$$P_{dev}(s) \coloneqq 3 \cdot \left(\left| I_{rot_t}(s) \right| \right)^2 \cdot R_{rot_t} \cdot \frac{(1-s)}{s}$$
$$P_{mech}(s) \coloneqq P_{dev}(s) - P_{mech_loss} \qquad \eta(s) \coloneqq \frac{P_{mech}(s)}{P_{sup}(s)}$$

 $s := 0.1 \cdot \%$, 0.2 %... $100 \cdot \%$



Figure 24 Mechanical output power versus slip.

5. Motor Speed

$$rpm := \frac{1}{\min} \qquad n_{sy} := \frac{f}{\frac{p}{2}}$$
$$n_m(s) := n_{sy} \cdot (1 - s) \qquad \omega_m(s) := 2 \cdot \pi \cdot n_m(s)$$

6. Motor Torque

$$T_{m}(s) := \frac{P_{mech}(s)}{\omega_{m}(s)}$$

s := 0.5%, 0.6% ... 80%



Figure 25 Torque versus slip.

 $s := 0.5\%, 0.6\% \dots 80\%$



Figure 26 Torque versus speed.



Motor Starting torque

- When the motor starts at s = 1,
- The ventilation losses are zero and the friction loss is passive. The negative friction loss does not drive the motor backwards.
- The mechanical losses are zero when s = 1
- This implies that the starting torque is calculated from the developed power instead of the mechanical output power.

Motor Starting torque

$$\mathbf{M}_{\mathrm{m}} \underbrace{\operatorname{start}(s) \coloneqq}_{2 \cdot \pi \cdot \mathrm{n}_{\mathrm{sy}} \cdot (1-s)} \frac{3 \cdot \left(\left| \mathbf{I}_{\mathrm{rot}}(s) \right| \right)^{2} \cdot \mathbf{R}_{\mathrm{rot}}(s) \cdot \frac{(1-s)}{s}}{2 \cdot \pi \cdot \mathrm{n}_{\mathrm{sy}} \cdot (1-s)}$$

$$M_{m _start}(s) := \frac{3 \cdot \left(\left| I_{rot_t}(s) \right| \right)^2 \cdot \frac{R_{rot_t}}{s}}{2 \cdot \pi \cdot n_{sy}}$$

• Kloss formula

$$M = \frac{2M_{MAX}}{\frac{S_{MAX}}{S} + \frac{S}{S_{MAX}}}$$

Circular diagram



- The resistances and reactance in the equivalent circuit for an induction motor can be determined by a series of measurements. The measurements are:
 - <u>No-load test</u>. This test determines the magnetizing reactance and core loss resistance.
 - <u>Blocked-rotor test</u>. This test gives the combined value of the stator and rotor resistance and reactance.
 - <u>Stator resistance measurement</u>.

No-load test

- The motor shaft is free
- The rated voltage supplies the motor.
- In the case of a three-phase motor:
 - the line-to-line voltages,
 - line currents
 - three-phase power using two wattmeters are measured

No-load test

Fig 29 Equivalent motor circuit in noload test



Fig. 30 Simplified equivalent motor circuit in no-load test

$$V_{no_load_ln} \coloneqq \frac{V_{no_load}}{\sqrt{3}} \qquad R_{c} \coloneqq \frac{V_{no_load_ln}}{P_{no_load_A}}$$

$$P_{no_load_A} \coloneqq \frac{P_{no_load}}{3} \qquad S_{no_load_A} \coloneqq V_{no_load_ln} \cdot I_{no_load}$$

$$Q_{no_load_A} \coloneqq \sqrt{S_{no_load_A}^2 - P_{no_load_A}^2}$$

$$X_{m} \coloneqq \frac{V_{no_load_ln}}{Q_{no_load_A}}$$

Blocked-rotor test

- The rotor is blocked to prevent rotation
- The supply voltage is reduced until the motor current is around the rated value.
- The motor is supplied by reduced voltage and reduced frequency. The supply frequency is typically 15 Hz.
- In the case of a three-phase motor:
 - the line-to-line voltages,
 - line currents
 - three-phase power using two wattmeters are measured

Blocked-Rotor test

Figure 31 Equivalent motor circuit for blocked-rotor test



Figure 32 Simplified equivalent motor circuit for blocked-rotor test



Blocked-Rotor test



The stator resistance was measured directly

$$R_{rot_t} := R_e - R_{sta}$$

Blocked-Rotor test

The magnitude of the motor impedance	$Z_{blocked} := \frac{V_{blocked_ln}}{I_{blocked}}$		
The leakage reactance at 15 Hz	$X_{e_{15Hz}} := \sqrt{Z_{blocked}^2 - R_e^2}$		
The leakage reactance at 60 Hz	$X_e := X_{e_15Hz} \cdot \frac{60Hz}{15Hz}$		

Numerical Example

A three -phase 30hp, 208V, 4 pole, 60Hz, wye connected induction motor was tested, the obtained results are:

No load test, 60 Hz	$V_{nL} := 208V$	$P_{nL} := 1600W$	$I_{nL} := 22A$	
Blocked Rotor test, 15Hz	V _{br} := 21V	$P_{br} := 2100W$	$I_{br} := 71A$	f _{br} := 15Hz
DC test	$V_{dc} := 12V$	I _{dc} := 75A		
Motor rating:	$P_{mot_rated} := 3$	Ohp V _{mot}	$V_{mot_ll} := 208V$	
	p := 4	f :=	60Hz	

Calculate :

a) The equivalent circuit parameters

b) Motor rated current and synchronous speed Draw the equivalent circuit

 $\underline{\text{No load test}}$, Determine the core losses Rc and magnetizing reactance Xm

Single phase values:

$$P_{nL_{1}F} := \frac{P_{nL}}{3} \qquad V_{nL_{nL}} := \frac{V_{nL}}{\sqrt{3}}$$

$$R_{c} := \frac{\left(V_{nL_ln}\right)^{2}}{P_{nL_1F}} \qquad R_{c} = 27.04\Omega$$

$$Y_{nL} := \frac{I_{nL}}{V_{nL_ln}} \qquad Y_{nL} = 0.183S$$

$$X_{m} := i \cdot \frac{1}{\sqrt{Y_{nL}^{2} - \left(\frac{1}{R_{c}}\right)^{2}}} \qquad X_{m} = 5.573i\Omega$$

Block Rotor test,

Neglect the magnetizing branch. Consider only the $X_{sta} + X_{rot}$ and $R_{sta} + R_{rot}$

$$X_{br} = X_{sta} + X_{rot}$$
 $R_{br} = R_{sta} + R_{rot}$

Single phase values:
$$P_{br_1F} := \frac{P_{br}}{3}$$
 $V_{br_ln} := \frac{V_{br}}{\sqrt{3}}$

Resistance value is:

$$R_{br} := \frac{P_{br_1F}}{I_{br}^2} \qquad \qquad R_{br} = 0.139\Omega$$

$$Z_{br} := \frac{V_{br} ln}{I_{br}} \qquad \qquad Z_{br} = 0.171\Omega$$

$$X_{br_15Hz} := i \cdot \sqrt{Z_{br}^2 - R_{br}^2}$$

 $X_{br_{15Hz}} = 0.099i\Omega$

The reactance at 60 Hz is:

$$X_{br_{60Hz}} := X_{br_{15Hz}} \cdot \frac{60}{15}$$
 $X_{br_{60Hz}} = 0.398i\Omega$

$$X_{br} := i \cdot \sqrt{\left(Z_{br}^{2} - R_{br}^{2}\right)} \cdot \frac{60 \text{Hz}}{f_{br}}$$
 $X_{br} = 0.398 \text{i}\Omega$
Induction Motors

Determination of R1 and R2 and X1 and X2

$$X_{sta} := \frac{X_{br}}{2} \qquad \qquad X_{rot} := X_{sta} \qquad \qquad X_{sta} = 0.199i\Omega$$

Y connected motor

Induction Motors

a) The equivalent circuit parameters

a) The equivalent circuit parameters



Induction Motors

C) Motor rated current and synchronous speed

